

Center for Quantum Networks

NSF Engineering Research Center

Error Correction for Quantum Networks

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https://cqn-erc.org/



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Part 1 Classical Error Correction By Michele Pacenti

Part 2 Quantum Error Correction By Narayanan Rengaswamy

Part 3 Application to Quantum Networks By Sijie Cheng









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T#ping is no substit\$te for th?nking.







Typing is no substitute for thinking.







Human languages are error correcting codes



 $S \setminus L$



Set of all the sounds humans can produce

Sounds not belonging to the language

Subset of sounds that form a language





Maximum likelihood decoding

İS no for



- → is
 - no
- - for
 - th?nking thinking





•Binary alphabet: operations over \mathbb{F}_2^n



•Each word is independent from the others (block codes)















Set (field) of all the binary vectors of length ${\cal N}$

Size of the space 2^n





Error correcting codes



Set (field) of all the binary vectors of length $\ensuremath{\mathcal{N}}$

• 2^k vectors $\in C$ • $(2^n - 2^k)$ vectors $\notin C$ $k \le n$







Codelength

Code dimension

010 •

 $\begin{array}{c} 0 \rightarrow 000 \\ 1 \rightarrow 111 \end{array} \hspace{1.5cm} \overset{\text{Codewords}}{\xrightarrow{}} \end{array}$

 $t = \lfloor \frac{d-1}{2} \rfloor$

Correctable errors



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Hamming distance = the number of positions in which two binary vectors differ







Binary Symmetric Channel (BSC)

- Each bit is flipped with probability lpha











X



















Maximum likelihood decoding









Maximum likelihood decoding



 $\hat{x} = \operatorname{argmax} \log_2 \overline{P(y|x)}$ $x \in C$

Maximum likelihood decoding



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d = d(x, y): the Hamming distance between x and y For the BSC channel: $P(y|x) = \alpha^d (1-\alpha)^{n-d}$ ML decoding rule:

$$\hat{x} = \underset{x \in C}{\operatorname{argmax}} P(x|y) = \underset{x \in C}{\operatorname{argmax}} \frac{P(x)P(y|x)}{P(y)}$$
$$\log_2 P(y|x) = d\log_2 \alpha + (n-d)\log_2(1-\alpha)$$
$$= d\log_2 \frac{\alpha}{1-\alpha} + n\log_2(1-\alpha)$$

negative for $\alpha < 1/2$

$$\hat{x} = \operatorname*{argmin}_{x \in \mathcal{C}} d(x, y)$$

Minimum distance decoding





- Let *u* be a binary vector of length *k*
- Let G be the generator matrix of the code C
- A codeword can be defined as $\chi = \iota \cdot G$

• Let *H* be the parity check matrix of the code *C* • \mathbf{x} is a codeword if and only if $\mathbf{x} \cdot \mathbf{H}^T = \mathbf{0}$



Codes as matrices



 $\mathbf{x} \in \operatorname{Im}\{\mathbf{G}\}$ $\mathbf{x} \in \operatorname{ker}\{\mathbf{H}\}$

$\mathbf{G} \cdot \mathbf{H}^T = \mathbf{0}$



$\mathbf{u} \in \{0, 1\}$ $\begin{cases} 0 \cdot \mathbf{G} = 000 \\ 1 \cdot \mathbf{G} = 111 \end{cases} \mathbf{G} = \begin{pmatrix} 1 & 1 & 1 \end{pmatrix} k$





Matrices of the [3,1,3] repetition code

 \mathcal{N} $\mathbf{H} = \begin{pmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \end{pmatrix} \begin{cases} n - k \end{cases}$ \mathcal{N}

 $111 \cdot \mathbf{H}^{T} = \begin{pmatrix} 1+1+0\\ 0+1+1 \end{pmatrix} = \begin{pmatrix} 0\\ 0 \end{pmatrix}$







- Assume we transmit a codeword *x* over a BSC
- The effect of the channel can be modeled as adding an error vector *e* to the codeword









 $\mathbf{s} = \mathbf{x} \cdot \mathbf{H}^T + \mathbf{e} \cdot \mathbf{H}^T$ $\mathbf{s} = \mathbf{e} \cdot \mathbf{H}^T$















Another example: the [7,4,3] Hamming code





C =

0000000,1110000,1001100,0111100,0101010,1011010,1100110,0010110,1101001,0011001,0100101,1010101,1000011,0110011,0001111,11111111.





Another example: the [7,4,3] Hamming code

0000000	1110000	1001100	0111100	0101010	1011010	1100110	0010110	1101001	0011001	0100101	1010101	1000011	0110011	0001111	1111111
1000000	0110000	0001100	1111100	1101010	0011010	0100110	1010110	0101001	101100 1	110010 1	001010 1	000001 1	111001 1	100111 1	0111111
0100000	1010000	1101100	0011100	0001010	1111010	1000110	0110110	1001001	011100 1	000010 1	111010 1	110001 1	001001 1	010111 1	1011111
0010000	1100000	1011100	0101100	0111010	1001010	1110110	0000110	1111001	000100 1	011010 1	100010 1	101001 1	010001 1	001111 1	1101111
0001000	1111000	1000100	0110100	0100010	1010010	1101110	0011110	1100001	001000 1	010110 1	101110 1	100101 1	011101 1	000011 1	1110111
0000100	1110100	1001000	0111000	0101110	1011110	1100010	0010010	1101101	001110 1	010000 1	101000 1	100011 1	011011 1	000101 1	1111011
0000010	1110010	1001110	0111110	0101000	1011000	1100100	0010100	1101011	001101 1	010011 1	101011 1	100000 1	011000 1	000110 1	1111101
0000001	1110001	1001101	0111101	0101011	1011011	1100111	0010111	1101000	001100 0	010010 0	101010 0	100001 0	011001 0	000111 0	1111110

The size of the table scales exponentially









[3,2,2] Single Parity Check code

[3,1,3] Repetition code



$\mathbf{H}_R \cdot \mathbf{H}_S^T = \mathbf{0}$



[3,2,2] Single Parity Check code

Dual codes C, C^{\perp}





Low Density Parity Check (LDPC) codes

Sparse parity check matrix



Low complexity iterative decoding (on graphs)









v_1 v_2 v_3 $\mathbf{H} = \begin{pmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \end{pmatrix} \begin{pmatrix} c_1 \\ c_2 \end{pmatrix}$





Tanner graph



 $V = \{v_1, v_2, v_3\}$ Variable nodes $C = \{c_1, c_2\}$ Check nodes $E = \{(v_j, c_i) \mid \mathbf{H}_{i,j} = 1\}$

• **Bipartite graph**













Tanner graph







- Low complexity algorithms
- Operate on the Tanner graph
- variable nodes and check nodes



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• Each iteration involves an exchange of messages between

• If the estimated error matches the syndrome, the decoder succeeds, otherwise it continues for another iteration





Bit-Flipping algorithm

Initialization – variable-to-check messages



Check-to-variable messages

 $\nu_{j \to i}^{(k)} = \begin{cases} \nu_{j \to i}^{(k-1)} \oplus 1, \text{ if } \\ i \in \\ \nu_{j \to i}^{(k-1)}, \text{ otherwise} \end{cases}$

Variable-to-check messages





- $\nu_{i \to i}^{(0)} = y_j \in \{0, 1\}$ (received value from channel)
- $\mu_{i \to j}^{(k)} = \bigoplus_{j \in \mathcal{N}(i)} \nu_{j \to i}^{(k-1)} \text{ (number of unsatisfied checks)}$

if
$$\left(\sum_{i\in\mathcal{N}(j)}\mu_{i\to j}^{(k)}\right) > \beta$$















y = 0 1 0 0 1 0 1 0 0 0 0 1 0 1 0 0 1 1 0 0 0 0 s = 0 1 1 0 0 1 0 0 1 1 0 0 1 0



 $\beta = 1$





$$\mu_{i \to j}^{(1)} = \bigoplus_{j \in \mathcal{N}(i)} \nu_{j \to i}^{(0)}$$



Example 2200200101101201111111

 $\nu_{j \to i}^{(1)} = \begin{cases} \nu_{j \to i}^{(0)} \oplus 1, \text{ if } \left(\sum_{i \in \mathcal{N}(j)} \mu_{i \to j}^{(1)} \right) > \beta \\ \nu_{j \to i}^{(0)}, \text{ otherwise} \end{cases}$













$$\mu_{i \to j}^{(2)} = \bigoplus_{j \in \mathcal{N}(i)} \nu_{j \to i}^{(1)}$$











Example













Transmitted codeword *x*

Received sequence y

$$\hat{x_j} = \underset{x_j \in \{0,1\}}{\operatorname{argmax}} P(x_j | \mathbf{y}) = \underset{x_j \in \{0,1\}}{\operatorname{argmax}} x_j \in \{0,1\}$$

Posterior marginal distributions







 $\hat{\mathbf{x}} = \operatorname{argmax} P(\mathbf{x}|\mathbf{y})$ $\mathbf{x} \in C$

Posterior joint probability distribution

 $\sum_{x_1,...,x_{j-1},x_{j+1},...,x_n \in \{0,1\}^{n-1}} P(x|\mathbf{y})$



Belief propagation (or sum-product)







 $= \operatorname{argmax}$ $x_1 \in \{0,1\}$





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 $\hat{x}_1 = \operatorname{argmax} P(x_1|y) = \operatorname{argmax}$ $x_1 \in \{0, 1\}$

P(x|y) $x_2, x_3, x_4, x_5 \in \{0, 1\}^4$

$$= \underset{x_{1} \in \{0,1\}^{4}}{\operatorname{argmax}} W_{1} \cdot \underbrace{\left(\prod_{j=1}^{5} W_{j} \right) \mathbb{I}(c_{1} = 0) \mathbb{I}(c_{2} = 0)}_{x_{2},x_{3},x_{4},x_{5} \in \{0,1\}^{4}} \left(\prod_{j=1}^{5} W_{j} \right) \mathbb{I}(c_{1} = 0) \mathbb{I}(c_{2} = 0)$$

 $x_4, x_5 \in \{0, 1\}^2$

 $VV_4 \cdot VV_5$




Belief propagation (or sum-product)



$$\nu_{x \to f}(x) = \prod_{h \in \mathcal{N}(x) \setminus \{f\}} \mu_{h \to x}(x)$$

$$u_{f \to x}(x) = \sum_{\{x\}} \left(f(X) \prod_{h \in \mathcal{N}(f) \setminus \{x\}} \nu_{h \to f} \right)$$

$$\sigma_i(x_i) = \prod_{h \in \mathcal{N}(x_i)} \mu_{h \to x_i}(x_i)$$

 $\mathcal{N}(f) \setminus \{x\}$







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Repeated random sampling to obtain numerical results

 MC simulation is the standard method for assessing the performance of a code/decoder

 Simulation software needs to be fast to generate the adequate number of samples in a relatively short time







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NSF-ERC Monte Carlo simulations of error correction

- Choose an error model
- Choose a code and a decoder
- Fix the number of samples (large enough)
- Fix the parameters of the error model (e.g., for BSC)
- For each sample:
 - Generate a codeword (usually the all-zero codeword is sufficient)
 - Generate a random sample of noise, add it to the codeword
 - Compute the syndrome and start decoding
 - If the decoder fails (doesn't converge or estimates a different codeword) \rightarrow count a failure
- Divide the number of failures for the total number of samples • Repeat for different parameters of the error model
- Plot error parameters vs failure rates













http://www.jaist.ac.jp/~kurkoski/teaching/portfolio/uec_s05/S05-LDPC%20Lecture%201.pdf

Plot examples









N. Raveendran, D. Declercq and B. Vasić, "A Sub-Graph Expansion-Contraction Method for Error Floor Computation," in IEEE Transactions on Communications, vol. 68, no. 7, pp. 3984-3995, 2020 55

Plot examples







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What do we gain from QEC?

Encode into a Quantum Error Correcting Code

ERRORS



United We Stand, Divided We Fall!



ime to failure







Universal gates on k logical qubits:









k qubits $|\psi\rangle_L$ QECC Encode n qubits



QECC: Quantum Error Correcting Code





Pauli Operators and Bloch Sphere



Phase Flip

Z

Z

Ζ

Bit-Phase Flip









The "Quantum" Binary Symmetric Channel:

$$\psi\rangle = \alpha |0\rangle + \beta |1\rangle - \epsilon(I, X, Z,$$

Depolarizing Channel





 $|0\rangle - Y - i |1\rangle$ $|0\rangle$ $|1\rangle - Y - i|0\rangle$ $|1\rangle - Z - |1\rangle$

Y)

 $|\psi\rangle$ $|\rho\rangle = \begin{cases} X |\psi\rangle \\ Z |\psi\rangle \end{cases}$ $Y |\psi\rangle$

with prob. $1 - \epsilon$, with prob. $\frac{\epsilon}{3}$, with prob. $\frac{\epsilon}{3}$, with prob. $\frac{\epsilon}{3}$.













- Pauli Group on *n* qubits:
- All possible tensor products of I, X, Z, Y with global phase from the set $\{1, i, -1, -i\}$
- 1. Group generated by I_i, X_i, Z_j, Y_i on j^{th} qubit
- 2. The 4ⁿ Hermitian operators form an orthogonal basis for all $2^n \times 2^n$ matrices
- 3. Trace Inner Product: $\langle A, B \rangle := Tr(A^{\dagger}B)$





Commutativity of Pauli Operators



n = 2 qubits Property: $(A \otimes B)(C \otimes D) = (AC) \otimes (BD)$















n = 2 qubits



Stabilizer Group: $S = \langle X \otimes X, Z \otimes Z \rangle = \{II, XX, ZZ, -YY\}$





Linear Algebra: When Hermitian matrices commute, they can be simultaneously diagonalized, i.e., they have a <u>common basis of eigenvectors</u>





Stabilizer Group: $S = \langle X \otimes X, Z \otimes X \rangle$

"Bell Basis": The common eigenbasis of these "Bell state" stabilizers

$$|\Phi^{+}\rangle = \frac{|00\rangle + |11\rangle}{\sqrt{2}} , \quad |\Phi^{-}\rangle = \frac{|00\rangle - |11\rangle}{\sqrt{2}} , \quad |\Psi^{+}\rangle = \frac{|01\rangle + |10\rangle}{\sqrt{2}} , \quad |\Psi^{-}\rangle = \frac{|01\rangle - |10\rangle}{\sqrt{2}}$$

$$A \otimes B(C \otimes D) = (AC) \otimes (BD) \Rightarrow XX |\Phi^{+}\rangle = \frac{X |0\rangle X |0\rangle + X |1\rangle X |1\rangle}{\sqrt{2}} = \frac{|11\rangle + |00\rangle}{\sqrt{2}} = (+1) |\Phi^{+}\rangle$$

Stabilizer Code Q(S): Subspace spanned by the +1 eigenvectors of S

In this case we only get a 1-dimensional space spanned just by $|\Phi^+\rangle$





Stabilizer Codes from Stabilizer Groups

$$\langle Z \rangle = \{II, XX, ZZ, -YY\}$$



Stabilizer Group: $S = \langle X \otimes X, Z \otimes X \rangle$

For *n* qubits, remove some of the *n* generators to make space for qubits!

$$|\Phi^+\rangle = \frac{|00\rangle + |11\rangle}{\sqrt{2}} \quad , \quad |\Phi^-\rangle = \frac{|00\rangle - |11}{\sqrt{2}}$$

Stabilizer Code Q(S): Subspace spanned by the +1 eigenvectors of S

$$\begin{split} |\psi\rangle &= \alpha |0\rangle + \beta |1\rangle \stackrel{\text{Encoding}}{\longrightarrow} \overline{|\psi\rangle} = \alpha |\Phi^+\rangle + \beta |\Phi^-\rangle \\ k &= 1 \text{ logical qubit} \qquad \qquad \equiv \alpha |00\rangle + \beta |11\rangle \\ n &= 2 \text{ physical qubits} \end{split}$$





Stabilizer Codes from Stabilizer Groups

$$\Diamond Z \rangle \to \langle Z \otimes Z \rangle \equiv \langle ZZ \rangle$$





Stabilizer Group: $S = \langle X \otimes X, Z \otimes Z \rangle \rightarrow \langle Z \otimes Z \rangle \equiv \langle ZZ \rangle$

logical qubit

Logical Pauli **Operators**

$$X |\psi\rangle = \alpha |1\rangle + \beta |0\rangle \leftarrow Z |\psi\rangle = \alpha |0\rangle - \beta |1\rangle \leftarrow C$$

Synthesis of Logical Operators: Translate a <u>universal</u> set of quantum gates







Stabilizer Codes: Error Correction

Stabilizer Group: $S = \langle Z \otimes Z \rangle \equiv \langle ZZ \rangle$; Logicals: $\overline{X} = XX, \overline{Z} = ZI \equiv IZ$

$$\begin{array}{c|c} k = 1 & |\psi\rangle = \alpha \left|0\right\rangle + \beta \left|1\right\rangle & \longrightarrow \\ \hline \text{ogical qubit} & |0\rangle & \longrightarrow \end{array} \right\} \overline{|\psi\rangle} = \alpha \left|00\right\rangle + \beta \left|11\right\rangle & \text{physical qubit} \end{array}$$

 \Rightarrow Suffices to consider only Pauli errors on the physical qubits*

- 1. Error E = XI occurs on the code states the states of the code states of the code states of the - 2. Measure ZZ, i.e., the property that the state belongs to the code





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Recall: Pauli operators form an orthogonal basis for all square matrices

te:
$$\overline{|\psi\rangle} \mapsto E\overline{|\psi\rangle} = \alpha |10\rangle + \beta |01\rangle$$

3. Since "syndrome" is -1, decoder estimates error to be $\hat{E} = XI$ or $\hat{E}' = IX = \hat{E}\bar{X}$





Stabilizer Codes: Error Correction

Stabilizer Group: $S = \langle Z \otimes Z \rangle \equiv \langle ZZ \rangle$; Logicals: $\overline{X} = XX, \overline{Z} = ZI \equiv IZ$

$$\begin{aligned} |\psi\rangle &= \alpha \left|0\right\rangle + \beta \left|1\right\rangle - \underbrace{-X}_{\left|0\right\rangle} \\ &= \underbrace{-X}_{\left|\psi\right\rangle} = \alpha \left|10\right\rangle + \beta \left|01\right\rangle \end{aligned}$$

- Measure Syndrome: Eigenvalues of $E|\psi\rangle$ with respect to stabilizer(s) $(ZZ)E\overline{|\psi\rangle} = (ZZ)(XI)\overline{|\psi\rangle} = (-1)(XI)(ZZ)\overline{|\psi\rangle} = (-1)E\overline{|\psi\rangle}$

 $\hat{E} = XI : \hat{E}E\overline{|\psi\rangle} = (XI)(XI)\overline{|\psi\rangle} = \overline{|\psi\rangle} \quad \hat{E} = IX : \hat{E}E\overline{|\psi\rangle} = (IX)(XI)\overline{|\psi\rangle} = \overline{X}\overline{|\psi\rangle}$ Logical Error!! **Correct Decoding**



Decoder: Takes syndrome as input and outputs most likely (Pauli) error \hat{E}











- Clearly, this is a trivial code that cannot store much information (i.e., logical qubits) or correct many errors
- We need a systematic approach to design good quantum codes and efficient decoders to correct errors on them
 - There is a rich literature on classical codes and decoders
- How can we establish a connection between quantum and classical codes to leverage that?





Bridge to Classical Codes

Map an *n*-qubit Hermitian Pauli matrix to a pair of binary vectors:

How to check if $E(a, b) = X \otimes Z \otimes Y$ and $E(c, d) = Z \otimes Z \otimes X$ commute? Compare operators on each qubit: $X \otimes Z \otimes Y \mapsto (\begin{bmatrix} 1 & 0 & 1 \end{bmatrix}, \begin{bmatrix} 0 & 1 & 1 \end{bmatrix})$ $Z \otimes Z \otimes X \mapsto (\begin{bmatrix} 0 & 0 \\ 1 \end{bmatrix}, \begin{bmatrix} 1 & 1 \\ 0 \end{bmatrix})$



Example for n = 3: $X \otimes Z \otimes Y \longrightarrow E(a, b)$ $a = \begin{bmatrix} 1 & 0 & 1 \end{bmatrix}$ (X component) $b = [0 \ 1 \ 1]$ (*Z* component)





Commutativity: Symplectic Inner Product





- How to check if $E(a, b) = X \otimes Z \otimes Y$ and $E(c, d) = Z \otimes Z \otimes X$ commute?
 - Compare operators on each qubit: $X \otimes Z \otimes Y \mapsto (\begin{bmatrix} 1 & 0 & 1 \end{bmatrix}, \begin{bmatrix} 0 & 1 & 1 \end{bmatrix})$ $Z \otimes Z \otimes X \mapsto (\begin{bmatrix} 0 & 0 & 1 \end{bmatrix}, \begin{bmatrix} 1 & 1 & 0 \end{bmatrix})$
 - Symplectic inner product: $\langle [a, b], [c, d] \rangle_{sym} \coloneqq ad^T + bc^T$ (modulo 2) $= \begin{cases} 0 & \text{iff they commute,} \\ 1 & \text{iff they anticommute} \end{cases}$



Commutativity: Symplectic Inner Product

= 1 + 1





- How to check if $E(a, b) = X \otimes Z \otimes Y$ and $E(c, d) = Z \otimes Z \otimes X$ commute?
 - Compare operators on each qubit: $X \otimes Z \otimes Y \mapsto (\begin{bmatrix} 1 & 0 & 1 \end{bmatrix}, \begin{bmatrix} 0 & 1 & 1 \end{bmatrix})$ $Z \otimes Z \otimes X \mapsto (\begin{bmatrix} 0 & 0 & 1 \end{bmatrix}, \begin{bmatrix} 1 & 1 & 0 \end{bmatrix})$ $ad^{T} + bc^{T} = \begin{bmatrix} 1 & 0 & 1 \end{bmatrix} \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix} + \begin{bmatrix} 0 & 1 & 1 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \end{bmatrix}$
 - $= 0 \pmod{2} \Rightarrow$ they commute





Recall: Syndrome is the commutation relations of error E(c, d) with stabilizers $(ZZ)E\overline{|\psi\rangle} = (ZZ)(XI)\overline{|\psi\rangle} = (-1)(XI)(ZZ)\overline{|\psi\rangle} = (-1)E\overline{|\psi\rangle}$ Syndrome = $\begin{bmatrix} \langle [\boldsymbol{a_1}, \boldsymbol{b_1}], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \\ \langle [\boldsymbol{a_2}, \boldsymbol{b_2}], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \end{bmatrix} = H_a \boldsymbol{d}^T + H_b \boldsymbol{c}^T \pmod{2}$



Quantum Parity-Check Matrices

$$\begin{array}{cccc} Z \otimes I & I \otimes Z \otimes Z \\ 0 & 0 & a_2 = \begin{bmatrix} 0 & 0 & 0 \end{bmatrix} \\ 1 & 0 & b_2 = \begin{bmatrix} 0 & 1 & 1 \end{bmatrix} \\ \end{array}$$

$$= \begin{bmatrix} H_a & H_b \end{bmatrix}; \quad \begin{array}{c} H_a H_b^T + H_b H_a^T = \mathbf{0} \\ \end{array}$$







$$|\Phi^+\rangle = \frac{|00\rangle + |11\rangle}{\sqrt{2}} , \quad |\Phi^-\rangle = \frac{|00\rangle - |11\rangle}{\sqrt{2}}$$

Stabilizers: $S = \langle XX, ZZ \rangle \rightarrow \langle ZZ \rangle$

 $|\psi\rangle = \alpha |0\rangle + \beta |1\rangle \xrightarrow{\text{Encoding}} \overline{|\psi\rangle} = \alpha |\Phi^+\rangle + \beta |\Phi^-\rangle$ $\equiv \alpha |00\rangle + \beta |11\rangle$

Encoding $\begin{aligned} |\psi\rangle &= \alpha \left|0\right\rangle + \beta \left|1\right\rangle \underbrace{\qquad }_{\left|0\right\rangle} \underbrace{\qquad }_{\left|\psi\right\rangle} &= \alpha \left|00\right\rangle + \beta \left|11\right\rangle \end{aligned}$

Logicals: $\overline{X} = XX, \overline{Z} = ZI \equiv IZ$



Bell, GHZ States \rightarrow 2- and 3-Qubit Codes

$$|\Phi_3^+\rangle = \frac{|000\rangle + |111\rangle}{\sqrt{2}} , \ |\Phi_3^-\rangle = \frac{|000\rangle - |11}{\sqrt{2}}$$

$S = \langle XXX, ZZI, IZZ \rangle \rightarrow \langle ZZI, IZZ \rangle$

$$|\psi\rangle = \alpha |0\rangle + \beta |1\rangle \xrightarrow{\text{Encoding}} \overline{|\psi\rangle} = \alpha |\Phi_3^+\rangle + \beta | \\ \equiv \alpha |000\rangle + \beta$$



 $\overline{X} = XXX, \overline{Z} = ZII \equiv IZI \equiv IIZ$











3-Qubit Code ≠ Quantum Repetition Code

$$\begin{aligned} |\psi\rangle &= \alpha |0\rangle + \beta |1\rangle \xrightarrow{\text{Encoding}} \overline{|\psi\rangle} &= \alpha |\Phi_3^+\rangle \\ &\equiv \alpha |000\rangle \\ &\text{Encoding} \\ |\psi\rangle &= \alpha |0\rangle + \beta |1\rangle \xrightarrow{\bullet} \\ &|0\rangle \xrightarrow{\bullet} \\ &|0\rangle \xrightarrow{\bullet} \\ &|\psi\rangle &= \alpha |00\rangle \end{aligned}$$

Recall Classical Repetition Code: $C = \{000, 111\}$ Quantum "No Cloning" Theorem: Arbitrary quantum state can't be cloned!



 $+\beta |\Phi_3^-\rangle$ $+\beta |111\rangle$

 $|00\rangle + \beta |111\rangle$

This is NOT the same as cloning (i.e., copying) the logical qubit into 3 physical qubits!!





Syndrome Measurement for 3-Qubit Code

Measure the stabilizer generators $S_1 = ZZI$ and $S_2 = IZZ$:





The error X propagates through the CNOT and flips the measurement

Hence, the measurement results in -1 whenever there are an odd number of X's on the ancilla (through the CNOT gates), i.e., when the error anticommutes with the stabilizer S_i (syndrome!)





Syndrome Measurement for 3-Qubit Code

Measure the stabilizer generators $S_1 = ZZI$ and $S_2 = IZZ$:





The error X propagates through the CNOT and flips the measurement

 $E(\boldsymbol{c},\boldsymbol{d}) = X_1 = XII$





Minimum Distance of 3-Qubit Code

Measure the stabilizer generators $S_1 = ZZI$ and $S_2 = IZZ$:



Minimum Distance of a Code: The minimum weight of an undetectable error = The minimum weight of a logical operator

Logicals: $\overline{X} = XXX$, $\overline{Z} = ZII \equiv IZI \equiv IIZ \Rightarrow$ Minimum Distance d = 1



The error Z DOES **NOT** propagate through the CNOT, so doesn't flip the measurement





Concatenate the 3-qubit code with its X-type version! This leads to the famous 9-qubit code of Peter Shor, called the "Shor code", which was the first time QEC was shown to work!





Clearly, this code can only correct 1 error of X-type and cannot even detect 1 error of Z-type

How can we construct a non-trivial code out of this 3-qubit code that can correct 1 error of any type?









Shor's 9-Qubit Code $S = \langle ZZI, IZZ \rangle, \overline{X} = XXX, \overline{Z} = ZII$ 3 Inner Each of the 3 physical qubits of Codes the initial 3-qubit "phase-flip" code is further encoded into the 6 3-qubit "bit-flip" code



Outer Concatenation

ΔΑΙ,ΙΔΑ $\overline{X} = XII, \overline{Z} = ZZZ$



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This is a [n = 9, k = 1, d = 3]code since the minimum weight of a logical operator is d = 3

There are 8 stabilizer generators, 6 of Z-type and 2 of X-type







[7,4,3] Hamming Code: $H = \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}$

[[7,1,3]] Steane Code: $H_S = \begin{bmatrix} I \end{bmatrix}$

 $H_a H_b^T +$ **Commutativity Check:**

Logical Pauli Operators:



Steane's 7-Qubit Code

$$H_a \mid H_b] = \begin{bmatrix} H \mid 0 \\ 0 \mid H \end{bmatrix} \begin{bmatrix} X \\ Z \end{bmatrix}$$

$$H_b H_a^T = \begin{bmatrix} 0 & HH^T \\ \hline 0 & 0 \end{bmatrix} = 0$$





Steane's

[7,4,3] Hamming Code: H =

[[7,1,3]] Steane Code: $H_S = | I$

Logical Pauli Operators:

 $\bar{X} = \begin{bmatrix} 1 \\ \bar{Z} = \begin{bmatrix} 0 \end{bmatrix}$

Since stabilizers fix code state, \overline{X} $Z \equiv (ZZZZZZZ)(Z_{3}Z_{5}Z_{6}Z_{7}) = Z_{1}Z_{2}Z_{4}$



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Decoding Steane's 7-Qubit Code

[[7,1,3]] Steane Code: $H_S =$

Error E(c, d): Syndrome = $H_a d^T$ +

This reduces to decoding the Hamming code for either type of error!

If X- and Z-errors are correlated, then decoders aren't independent





$$H_a \mid H_b] = \begin{bmatrix} H \mid 0 \\ 0 \mid H \end{bmatrix} \begin{bmatrix} X \\ Z \end{bmatrix}$$

$$H_b \boldsymbol{c}^T = \begin{bmatrix} H \\ 0 \end{bmatrix} \boldsymbol{d}^T + \begin{bmatrix} 0 \\ H \end{bmatrix} \boldsymbol{c}^T = \begin{bmatrix} H \boldsymbol{d}^T \\ H \boldsymbol{c}^T \end{bmatrix} = \begin{bmatrix} \boldsymbol{s}_X \\ \boldsymbol{s}_Z \end{bmatrix}$$

Decoding: Use the X-syndrome s_X (Z-syndrome s_Z) to estimate Z-error (X-error)







Recall Standard Array Decoding

0000000	1110000	1001100	0111100	0101010	1011010	1100110	0010110	1101001	0011001	0100101	1010101	1000011	0110011	0001111	1111111
1000000	0110000	0001100	1111100	1101010	0011010	0100110	1010110	0101001	1011001	1100101	0010101	0000011	1110011	1001111	0111111
0100000	1010000	1101100	0011100	0001010	1111010	1000110	0110110	1001001	0111001	0000101	1110101	1100011	0010011	0101111	1011111
0010000	1100000	1011100	0101100	0111010	1001010	1110110	0000110	1111001	0001001	0110101	1000101	1010011	0100011	0011111	1101111
0001000	1111000	1000100	0110100	0100010	1010010	1101110	0011110	1100001	0010001	0101101	1011101	1001011	0111011	0000111	1110111
0000100	1110100	1001000	0111000	0101110	1011110	1100010	0010010	1101101	0011101	0100001	1010001	1000111	0110111	0001011	1111011
0000010	1110010	1001110	0111110	0101000	1011000	1100100	0010100	1101011	0011011	0100111	1010111	1000001	0110001	0001101	1111101
0000001	1110001	1001101	0111101	0101011	1011011	1100111	0010111	1101000	0011000	0100100	1010100	1000010	0110010	0001110	1111110



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[[7,1,3]] Steane Code: $H_S = \begin{bmatrix} H_a & H_b \end{bmatrix} = \begin{bmatrix} H & 0 \\ 0 & H \end{bmatrix} \begin{bmatrix} X \\ Z \end{bmatrix}$







Calderbank-Shor-Steane (CSS) Codes

[[7,1,3]] Steane Code $H_{S} = \begin{bmatrix} H & 0 \\ 0 & H \end{bmatrix} \begin{array}{c} X \\ Z \end{array}$ $HH^{T} = 0$

Syndrome =

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$$\begin{bmatrix} H \boldsymbol{d}^T \\ H \boldsymbol{c}^T \end{bmatrix} = \begin{bmatrix} \boldsymbol{s}_X \\ \boldsymbol{s}_Z \end{bmatrix}$$

 $\overline{X} = E(\boldsymbol{c}, \boldsymbol{0}) ; H\boldsymbol{c}^{T} = \boldsymbol{0}$ $\overline{Z} = E(\boldsymbol{0}, \boldsymbol{d}) ; H\boldsymbol{d}^{T} = \boldsymbol{0}$

 $k = n - \operatorname{rank}(H_X) - \operatorname{rank}(H_Z)$



[[n, k, d]] CSS Code

$H_{S} = \begin{bmatrix} H_{X} & 0 \\ 0 & H_{Z} \end{bmatrix} \begin{bmatrix} X \\ Z \end{bmatrix}$ $H_{X}H_{Z}^{T} = 0$ $\begin{bmatrix} H_{X}d^{T} \end{bmatrix} \begin{bmatrix} s_{X} \end{bmatrix}$

Syndrome = $\begin{bmatrix} H_X d^T \\ H_Z c^T \end{bmatrix}$ = $\begin{bmatrix} s_X \\ s_Z \end{bmatrix}$

 $\overline{X} = E(\boldsymbol{c}, \boldsymbol{0}) ; \ H_Z \boldsymbol{c}^T = \boldsymbol{0}$ $\overline{Z} = E(\boldsymbol{0}, \boldsymbol{d}) ; \ H_X \boldsymbol{d}^T = \boldsymbol{0}$

 $d = \text{minimum weight of } \overline{X}, \overline{Z}$




CSS Quantum LDPC (QLDPC) Codes [[n, k, d]] CSS Code , d_X] $H_S = \begin{bmatrix} H_X & 0 \\ 0 & H_Z \end{bmatrix} \begin{bmatrix} X \\ Z \end{bmatrix}$ icals H_X/H_Z $H_X H_Z^T = 0$ Syndrome = $\begin{bmatrix} H_X d^T \\ H_Z c^T \end{bmatrix} = \begin{bmatrix} s_X \\ s_Z \end{bmatrix}$ lizers $\overline{X} = E(\boldsymbol{c}, \boldsymbol{0}) ; \ H_Z \boldsymbol{c}^T = \boldsymbol{0}$ $\overline{Z} = E(\mathbf{0}, \mathbf{d}) ; H_X \mathbf{d}^T = \mathbf{0}$

$[n, k_Z, d_Z] \underset{I}{\mathcal{C}_Z} \longrightarrow LDPC \longleftarrow \underset{I}{\mathcal{C}_X} [n, k_X]$				
$X-logicals$ $\overline{X} \equiv G_Z/H_X$	$k = k_Z - (k_Z - k_Z)$ $= k_X - (k_Z - k_Z)$	$(n-k_X)$ $(n-k_Z)$	$\frac{Z - \log}{\overline{Z} \equiv G}$	
C_X^{\perp}		⊥ Z		
X-stabilizers H _X	$n-k_X$	$n-k_Z$	Z-stabil H _Z	
)	((



d = minimum weight of X, Z







The Surface Code



Errors cause syndrome at endpoints





Minimum Weight Perfect Matching

X-Syndrome Graph



Virtual Check for Matching



Errors cause syndrome at endpoints

 \Box - vertex checks (H_X) \Box - plaquette checks (H_Z)





MWPM Decoding Threshold



Wang et al. http://arxiv.org/abs/0905.0531 Errors cause syndrome at endpoints





 \Box - vertex checks (H_X) \Box - plaquette checks (H_Z)





Hypergraph Product (HP) QLDPC Codes [[n, k, d]] CSS Code [[N, K, d]] HP Code $H_S = \begin{bmatrix} H_X & 0 \\ 0 & H_Z \end{bmatrix} \begin{bmatrix} X \\ Z \end{bmatrix}$ $H_X H_Z^T = 0$ Syndrome = $\begin{bmatrix} H_X d^T \\ H_Z c^T \end{bmatrix} = \begin{bmatrix} s_X \\ s_Z \end{bmatrix}$

 $C_a: [n_a, k_a, d_a] \to H_a: m_a \times n_a$ $C_b: [n_b, k_b, d_b] \to H_b: m_b \times n_b$



 $N = n_a m_b + n_b m_a$ $\overline{K} = 2k_a k_b - k_a (n_b - m_b) - k_b (n_a - m_a) \sim \Theta(N)$ $d \geq \min(d_a, d_b, d_a^T, d_b^T) \sim \Theta(\sqrt{N})$



 $\overline{X} = E(\boldsymbol{c}, \boldsymbol{0}) ; \ H_Z \boldsymbol{c}^T = \boldsymbol{0}$ $\overline{Z} = E(\mathbf{0}, \mathbf{d}) ; H_X \mathbf{d}^T = \mathbf{0}$

d = minimum weight of X, Z





[[N, K, d]] HP Code

 $C_a: [5, 1, 5] \to H_a: m_a \times n_a = 4 \times 5$ $C_b: [4,0,0] \rightarrow H_b = H_a^T: m_b \times n_b = 5 \times 4$



$$N = n_a m_b + n_b m_a$$

$$K = 2k_a k_b - k_a (n_b - m_b) - k_b (n_a - m_a) =$$

$$d = 5$$







Belief propagation (or sum-product)







 $= \operatorname{argmax}$ $x_1 \in \{0,1\}$





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 $\hat{x}_1 = \operatorname{argmax} P(x_1|y) = \operatorname{argmax}$ $x_1 \in \{0, 1\}$

P(x|y) $x_2, x_3, x_4, x_5 \in \{0, 1\}^4$

$$= \underset{x_{1} \in \{0,1\}^{4}}{\operatorname{argmax}} W_{1} \cdot \underbrace{\sum_{x_{2},x_{3},x_{4},x_{5} \in \{0,1\}^{4}}_{x_{2},x_{3},x_{4},x_{5} \in \{0,1\}^{4}} \left(\prod_{j=1}^{5} W_{j}\right) \mathbb{I}(c_{1}=0)\mathbb{I}(c_{2}=0)$$

$$= \underset{x_{1} \in \{0,1\}}{\operatorname{argmax}} W_{1} \cdot \underbrace{\sum_{x_{2},x_{3} \in \{0,1\}^{2}} \mathbb{I}(c_{1}=0) \cdot W_{2} \cdot W_{3}}_{x_{1} \in \{0,1\}}$$

 $x_4, x_5 \in \{0, 1\}^2$

$$+x_3 = 0$$

 $+x_5 = 0$

 $V_4 \cdot V_5$





Iterative Decoding of QLDPC Codes

$[n, k_Z, d_Z]$ (C)	$Z \longrightarrow LD$	PC <i>C</i>	X [n, k_X
$X-logicals$ $\overline{X} \equiv G_Z/H_X$	$k = k_Z - k_Z - k_X - $	$(n-k_X)$ $(n-k_Z)$	$\frac{Z - \log}{\overline{Z} \equiv G}$
C_X^{\perp}		С	$\frac{\bot}{Z}$
X-stabilizers H _X	$n-k_X$	$n-k_Z$	Z-stabi H _Z
$\langle 0 \rangle$		$\langle 0 \rangle$	



, $d_X]$ cicals $_X/H_Z$

lizers

Syndrome = $\begin{bmatrix} H_X d^T \\ H_Z c^T \end{bmatrix}$ = $oldsymbol{s}_X$

- 1. Decoding is based on syndrome, no received vector
- 2. No X/Z Correlations in Error:

Run two independent iterative

decoders based on H_X , H_Z

3. <u>Depolarizing Channel:</u> Exchange information or run non-binary decoder for X, Z, Y

d = minimum weight of X, Z











Iterative Decoding at Finite Lengths



Raveendran, Vasić: <u>https://arxiv.org/abs/2012.15297</u>



Syndrome = $\begin{bmatrix} H_X \boldsymbol{d}^T \\ H_Z \boldsymbol{c}^T \end{bmatrix} =$ \boldsymbol{s}_X ${oldsymbol s}_Z$

- 1. <u>Decoding Schedule:</u> An iterative decoder can run parallel or serial node updates 2. Error Floor: The logical error rate saturates ("floors") once physical error rate is small
- 3. <u>Degeneracy:</u> Two errors differing by a stabilizer are equivalent, are "degenerate"







Recent Exciting QLDPC Codes

IEEE TRANSACTIONS ON INFORMATION THEORY, VOL. 60, NO. 2, FEBRUARY 2014

Quantum LDPC Codes With Positive Rate and Minimum Distance Proportional to the Square Root of the Blocklength

Jean-Pierre Tillich and Gilles Zémor

Degenerate Quantum LDPC Codes With Good Finite Length Performance

Pavel Panteleev and Gleb Kalachev

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IEEE TRANSACTIONS ON INFORMATION THEORY, VOL. 68, NO. 1, JANUARY 2022

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Quantum LDPC Codes With Almost Linear Minimum Distance

Pavel Panteleev[®] and Gleb Kalachev[®]

Fiber Bundle Codes: Breaking the $N^{1/2}$ polylog(*N*) **Barrier for Quantum LDPC Codes**

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IEEE TRANSACTIONS ON INFORMATION THEORY, VOL. 67, NO. 10, OCTOBER 2021

Balanced Product Quantum Codes

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Asymptotically Good Quantum and Locally Testable Classical **LDPC Codes**

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Quantum Low-Density Parity-Check Codes

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Quantum Tanner codes

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Towards "Good" Quantum LDPC Codes



Topological Codes	Optimal QLDPC Cod
$[[n, 1, \Theta(\sqrt{n})]]$	$[[n, \Theta(n), \Theta(n)]]$
High error thresholds	Promising threshold
~ Linear-time decoder	Linear-time decoder
Logical gates known	Very little research
Nearest-neighbor	Long-range interactions
Not scalable; large overhead	Scalable with constant overhead?



















Part 1 Classical Error Correction By Michele Pacenti

Part 2 Quantum Error Correction By Narayanan Rengaswamy

Part 3 Application to Quantum Networks By Sijie Cheng











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$$|\psi\rangle = \alpha |0\rangle + \beta |1\rangle$$

Set
$$\alpha = \beta = \frac{1}{\sqrt{2}}$$







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Bell State Measurement Property



 $|\psi\rangle = rac{|00\rangle + |11\rangle}{\sqrt{2}}$

















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Correct

Legend **Un-coded State with Error**

Coded State with Error

Syndrome for Error



Quantum Channel without Error

Quantum Channel with Error

Bell Pair without Error

Bell Pair with Error

Equality







- Un-coded State without Error + Un-coded State with Error **Stabilizer Generator Coded State without Error Estimated Error** Sign for Stabilizers







Quantum Network





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Scenario 2: Decoder failure



Problem 1 – Long Distance



1. As the distance increases, the quantum channel may experience more errors, potentially causing the decoder to fail.

2. If decoding is performed in realtime, a decoder failure will result in the quantum link being cut off. We need to find a method to perform decoding earlier, yet not in real-time, to reduce the impact of such failures. 130













Scenario 2: Decoder failure



Problem 2 – Quantum Circuit Error

1. As the distance increases, the quantum channel may experience more errors, potentially causing the decoder to fail.

2. If decoding is performed in realtime, a decoder failure will result in the quantum link being cut off. We need to find a method to perform decoding earlier, yet not in real-time, to reduce the impact of such failures.















Alice



Teleportation

Bob





Alice

Bob

Decoder















Quantum Error Correction



Teleportation













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Course Evaluation Survey

We value your feedback on all aspects of this short course. Please go to the link provided in the Zoom Chat or in the email you will soon receive to give your opinions of what worked and what could be improved.







